

Channel Model at 868 MHz for Wireless Sensor Networks in Outdoor Scenarios

J.M. Molina-Garcia-Pardo, A. Martinez-Sala, M.V. Bueno-Delgado, E. Egea-Lopez, L. Juan-Llacer, J. García-Haro.
 Department of Information Technologies and Communications,
 Polytechnic University of Cartagena, E-30202, Spain, Tel: +34 968325363, Fax: +34 968325973
 E-mail: {josemaria.molina, alejandros.martinez, mvictoria.bueno, esteban.egea, leandro.juan, joang.haro} @upct.es

Abstract—Wireless Sensor Networks (WSN) are formed by a large number of sensing nodes at the ground level. These devices are monitoring and measuring physical parameters from the environment. Simulation is used to study WSN, since deploying test-beds supposes a huge effort. However simulation results rely on physical layer assumptions, which are not usually accurate enough to capture the real behaviour of WSN. In this work several measurement campaigns are performed in three different scenarios: an open quasi-ideal area, a university yard and a park. The main contribution of this work is that a two slopes lognormal path-loss near ground outdoor channel model at 868 MHz is validated, and compared to the widely used one slope model. This model is useful for simulations because its computational cost is low.

Index Terms—Wireless Sensor Network, near ground measurement, log-normal path-loss, channel modeling.

I. INTRODUCTION

Wireless Sensor Networks (WSNs) are a new paradigm of telecommunication networks. WSNs are intended to allow efficient data collection and event control. WSNs share key properties with Mobile Ad-hoc NETWORKS (MANETs): Decentralized control, common transmission channel, broadcast nature, multi-hop routing and ephemeral topologies. However, unlike MANETs, WSNs must face: (a) specific traffic patterns, characterized by very long idle periods and sudden peak transmissions, (b) long run battery-powered deployment, that yields to tight energy constraints, and (c) device (hardware and software) simplicity. Therefore, two fundamental goals of WSN protocols are energy saving and traffic/environment adaptability. Therefore there is a challenge in developing energy-efficient protocols since nodes may not replace the battery and have to collaborate in a distributed manner to setup and self-organize the network [1].

Real applications are being explored and some of them are yet to come. Deploying and operating a test-bed to study the real behavior of protocols and network performance implies a great effort [2,3,4]. In [2,3,4] they perform the studies using experimental measurements with a widely used Mica2 Motes [5] and they all agree that the wireless links exhibit a random

and unreliable behavior. Also, WSN protocols are often developed and evaluated by means of simulations that make a simplistic approach of the radio layer.

As credibility of high level protocols simulation results depend on physical layer assumptions, non-realistic radio models can lead to mistaken results [6]. However, more realistic channel models add increasing computational requirements which cannot be usually afforded. This problem specially affects WSN with hundreds or thousands of nodes. In [7], Naomov and Gross show the scalability problems of ns-2 [8] working just with 100 nodes. A channel model for simulations should be accurate enough without increasing computational costs.

Therefore the goal of this paper is to achieve more accurate but still low computational near ground channel models to feed more realistic WSN simulations. In this paper we present the channel characterization for these networks in three outdoor scenarios. At each scenario we compare two models: One slope log-distance path loss model and two slope log-distance path loss model. The former is used as a ns-2 propagation model. We provide adjustment model parameters that can be used with ns-2 WSN simulations. The latter provides better accuracy with a similar computational cost, thus achieving both objectives of this paper.

A. Assumptions

A Typical Wireless Sensor Network application running in an outdoor scenario has been considered for this work [1]. Such an application may be a surveillance battlefield, a habitat monitoring, a disaster relief, or a location and navigation system for explorer robots in Mars, etc. These applications have in common that Motes/nodes are deployed randomly throughout the area. We also assume that in these networks the nodes are static and channel variations are due to the environment, which can be considered quasi-static (slow changing assumption).

The Motes/nodes have a low power narrow-band RF transceiver (as Mica2 Motes [9] working at 868 MHz (Europe) or 915 MHz (USA) in the free Industrial, Scientific and Medical (ISM) frequency bands with a maximum radiated power of 5 dBm. Our research has been conducted at 868 MHz; however, the propagation characteristics and conclusions can be also extended to 915MHz (frequency variation is very low, around 5%). The transmission rate for these systems is very low (a few dozens of Kbps) and the

symbol period much higher than RMS delay spread, therefore a flat slow fading channel can be assumed [10].

The rest of the paper is organized as follows: section II contains the related work. In section III the channel sounder equipment is explained. Section IV describes the measurement campaigns; different scenarios and the methodology are presented. Results are described in Section V. Finally, section VI presents the main conclusions.

II. RELATED WORK

As stated above, the nodes are over the ground and their antennas are a few centimeters over the ground. It should be remarked that there is a lack of near ground measurements [11] in scientific literature, and the vast majority of studies place antennas at a height greater than a meter [10]. In addition, there is an increasing interest in evaluating and measuring the actual link behaviour and its effect and influence for WSN but there is a lack of a channel characterisation that explains in a systematic way the propagation behaviour. In [12], Zhao and Govindan provide one of the first works that offered experimental measurements that shows that a wireless link is unreliable but they do not give any explanation of its findings. In [13] it is presented empirical measurements of the link quality that reflect the impact on routing protocols and it is shown the need for the implementation of a link quality estimator in a node. Recently, in [14] it is shown empirical results of extensive link layer measurements with the Eyes Nodes but they do not provide any channel model.

In [4] it is reported that the radio irregularity has a big influence on the routing protocols and they propose the Radio Irregularity Model that takes into account the non-isotropic properties of the antenna and the heterogeneous properties of node hardware.

III. CHANNEL SOUNDER EQUIPMENT

The channel sounder set-up is displayed in Fig. 1. A Mica2 Motes [5, 9] transceiver is used as transmitter and a spectrum analyzer (ROHDE & SCHWARZ FSH-3) is used as receiver, and both of them are placed just over the ground.

Taking into account that the goal of this work is to characterize the near ground wireless channel, but not the effects of the antennas, two $\lambda/4$ monopoles placed over two $\lambda/4 \times \lambda/4$ ground planes are employed, instead of using the Mica2 whip antennas. Both antennas are connected to the transmitter and to the receiver by a low-losses coaxial cable, and are located far from the hardware to minimize its influence (see Fig. 1).

The receiver consists of a laptop, a spectrum analyzer and batteries mounted in a woody trolley with absorber foam to reduce near scatter (see Fig. 1).

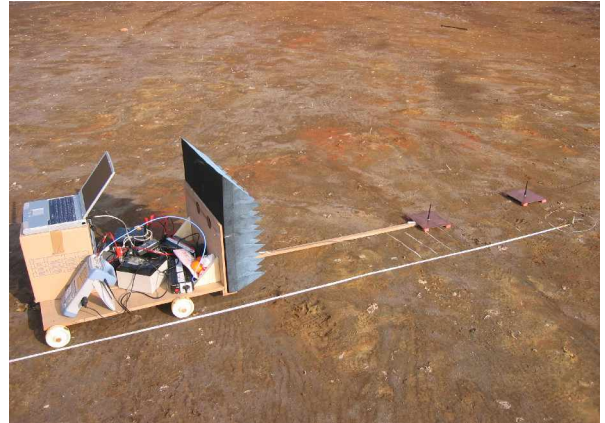


Fig 1: Photo of the channel sounder set-up in the ground plain scenario

The receiving antenna is placed at one and a half meter far away from the trolley, whereas the transmitting antenna is one meter far away from the Mica2 transceiver. The transmitter sends a constant carrier of 5 dBm (maximum radiated power) at 868.2972 MHz and the spectrum analyzer is set to this central frequency.

IV. MEASUREMENTS CAMPAIGNS

A. Scenarios

Three different outdoor environments are characterized in this work: A ground plain area, a university yard and a green park. The first one is a huge ground plain (quasi ideal flat ground) close to the sea without any important close scatter. In this environment propagation can be considered quasi-ideal. The second one is a university yard with dimensions 50m x 35m surrounded by four-story buildings with reinforced cement floors and walls. Here a lot of multipath components contribute at the receiver. Finally, the third of them is a grass green park slightly curved next to a five-story wall.

B. Methodology

For all scenarios, the same methodology is applied. Several random positions are chosen for the transmitter, and for each of them the trolley is separated from the transmitter until the received signal is 10 dB over the thermal noise (for this spectrum analyzer the measured noise is -95 dBm for a 10 KHz bandwidth). The steps between the receiver positions and the transmitter are: From 0.5 to 3 meters every 0.5 meters, from 3 to 10 meters every 1 meter and then every 2 meters.

Centered at each step (defined above), a space averaging at 5 positions separated 10 cm is done, and for all these five positions a 5 times time averaging is also performed (25 snapshots per step). The experiments are conducted several times at each scenario, and for each environment similar results are obtained in different runs. The slow time variance assumption was checked; it was observed that the time variability of the measurements in each position was negligible.

V. RESULTS

Fig. 2 shows the results for one run in the first scenario according to the methodology explained above (time averaged received power). It can be seen that the received power fits a straight line using the method of least square errors (log-distance attenuation) when the distance is expressed in dB [15]. The same behaviour is found in the other measurements and scenarios. Therefore, a log-normal path loss model may be considered (the normality is checked below). In addition, the antenna heights are very low (less than λ), and an important part of the Fresnel zone is always obstructed by the ground, so a two-segment least-square fit line may also be used. Then, at each scenario two models are compared, one slope log-distance path loss model and two slopes log-distance path loss model. These models can be denoted by expressions (1) and (2) [10].

$$L(d) = L_0 + 10n \log_{10}(d) + X_{\sigma} \quad (1)$$

$$L(d) = \begin{cases} L_{0_1} + 10n_1 \log_{10}(d) + X_{\sigma_1}, & d \leq d_r \\ L_{0_2} + 10n_2 \log_{10}(d) + X_{\sigma_2}, & d > d_r \end{cases} \quad (2)$$

Where in (1) L_0 is the path loss at 1 meter, n is the decay factor, d is the distance expressed in meters, and X_{σ} is a lognormal variable with standard deviation of σ (expressed in dB). In (2) two different slopes are defined before and after a breakpoint d_r . For each measured environment these models were applied to each run. It can be clearly observed that a two slopes model fits quite well with the measurements (Fig. 2). Also, the probability distributions of X_{σ} , X_{σ_1} and X_{σ_2} have been studied. Fig3, 4 and 5 shows the probability density functions of X_{σ} , X_{σ_1} and X_{σ_2} .

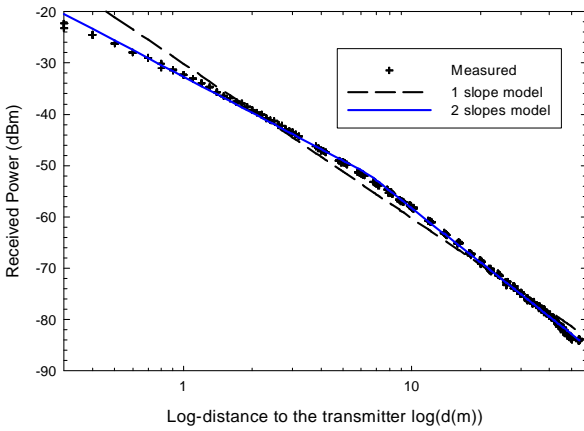


Fig 2: Results for the ground plain scenario

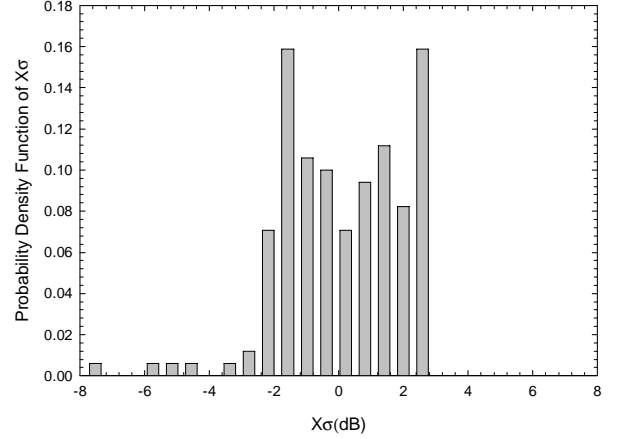


Fig 3. Probability Distribution Function of X_{σ} (one slope model)

It can be observed that X_{σ} does not follow a Gaussian distribution, whereas X_{σ_1} and X_{σ_2} do. An exhaustive measurements campaign would be needed to increase the resolution of these results.

The two high peaks of Fig. 3 are due to the mismatch of the one slope model with the measurements.

For all the cases under study, all the results were averaged and they are summarised in table I. It is observed that the breakpoint depends on the environment, and that it cannot be calculated as in [15], where it is said that it only depends on the antennas height and the carrier frequency (13 cm for this configuration). The park scenario was slightly curved, which explains that the breakpoint becomes even closer to the transmitter (a bigger Fresnel zone is intercepted by the ground).

Comparing both models, it is revealed that a two slopes model leads to a more accurate channel characterisation, and that the standard deviation decreases and its probability density function fits better into a Gaussian. In addition, in this model n_2 tends to 4, which is the expected value for a line of sight situation after the breakpoint. This parameter is crucial for a simulator, because it is going to fix the maximum transmission range of a node.

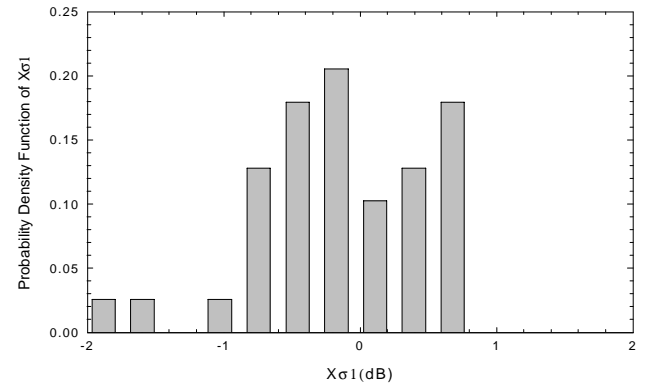


Fig 4: Probability Distribution Function of X_{σ_1}

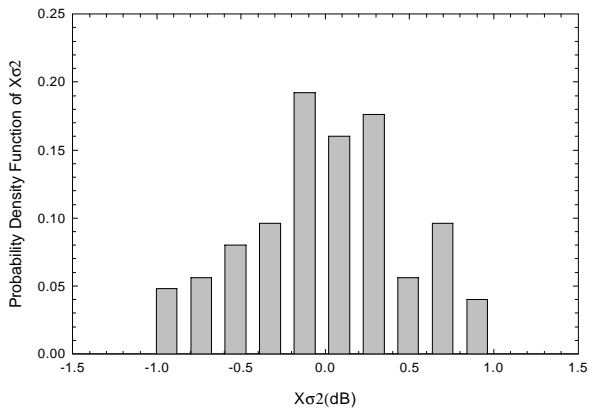


Fig 5: Probability Distribution Function of X_{σ_2}

In table I the maximum averaged radio coverage was calculated for a transmitted power of 5 dBm and a sensitivity of -100 dBm using both models. It is concluded that the two slopes model is more restrictive than the one slope model.

VI. CONCLUSIONS

Multiple research papers in the field have presented simulation results for Wireless Sensor Networks. The majority of these works rely on simplistic physical layer assumptions of channel models that can lead to mistaken results.

In this paper we present the results of several measurements campaigns that have been performed in order to obtain more accurate and low-computational near ground channel models to feed realistic simulation for the emergent WSN technology. This may contribute to obtain more reliable results and conclusions of protocol performance.

Three different scenarios/environments have been characterized and the same methodology has been applied. It has been validated a two slope lognormal channel model at 868 MHz that achieves a more accurate propagation characterization than those of one slope. It also achieves a more accurate propagation characterization than those of one slope, which do not follow a lognormal distribution. Model parameters for WSN propagation are provided. They can be used to adjust ns-2 propagation model for WSN simulation. The two-slope lognormal provides improved accuracy and does not increase computational cost, being suitable for large scale WSN simulations.

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TABLE I
ADJUSTMENT MODEL PARAMETERS FOR THE THREE SCENARIOS

	Symbol	Ground Plain	University Yard	Park
1 slope model	n	3.07	3.57	3.69
	σ	1.83	3.27	1.42
	d_{\max} (m)	189.5	41.3	52.4
2 Slopes Model	n_1	2.35	2.76	2.18
	n_2	3.6	4.01	3.95
	σ_1	0.6	2.98	0.28
	σ_2	0.42	1.82	0.67
	Breakpoint (m)	6.2	3.2	0.95
	d_{\max} (m)	139.8	32.4	45.7

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